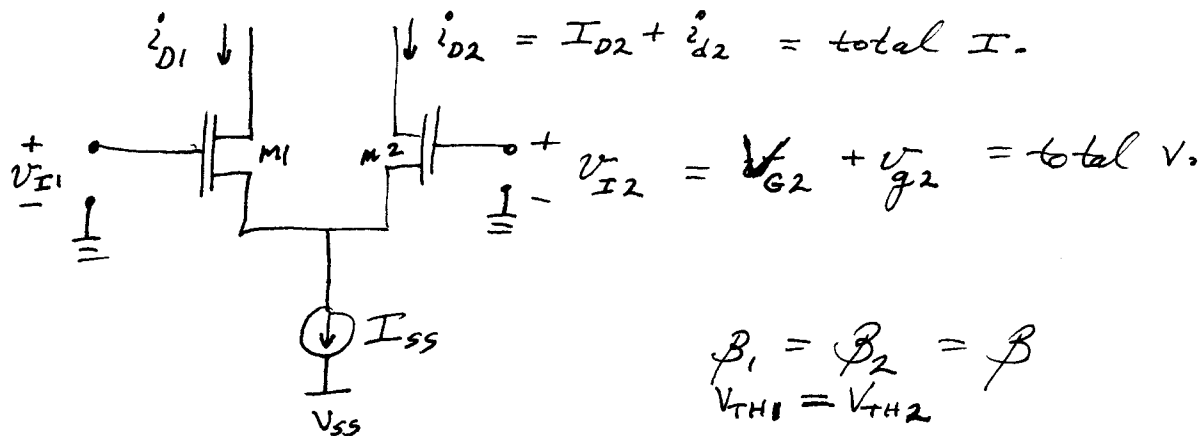


Differential Amps Diff-Amp24.1 SOURCE Coupled Pair

• Define $v_{DI} = v_{I1} - v_{I2} =$

$$= v_{G1} - v_{G2} = v_{G1} - v_{G1} - (v_{G2} - v_{G2})$$

$$= v_{GS1} - v_{GS2}$$

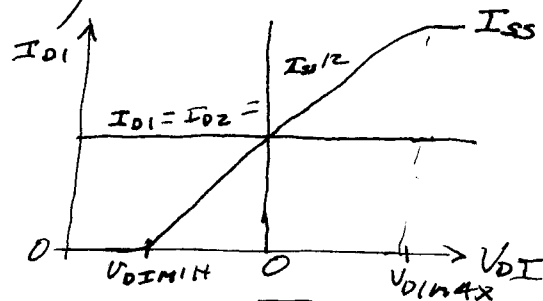
$$v_{DI} = \sqrt{\frac{2}{\beta}} (\sqrt{i_{D1}} - \sqrt{i_{D2}})$$

• GROUND $v_{I2} = 0$

• $\frac{v_{DI\max}}{v_{DI}} = v_{I1}$

• FIX v_{I2} & $\uparrow v_{I1}$ until $I_{SS} = I_{D1}$

$$v_{DI\max} = \sqrt{\frac{2I_{SS}}{\beta}}$$



$$\begin{cases} I_{D1} = I_{SS} \\ I_{D2} = 0 \end{cases}$$

M2 turning off
 $\Rightarrow v_S = v_{I2} - V_{THN}$

• $v_{DI\min}$

$\downarrow v_{I1}$ until M1 shuts off, $I_{D2} = I_{SS}$

$$v_S = v_{I1} - V_{THN}$$

$$v_{DI\min} = -v_{DI\max}$$

$$v_{GS2} = v_{I2} - v_{I1} + V_{THN} = -v_{DI\min} + V_{THN}$$

Differential Transconductance

$$G_m \equiv \left. \frac{di_{D1}}{dv_{DI}} \right|_{v_{DI}=0} = ?$$

$$v_{DI} = \sqrt{\frac{2}{\beta}} \left[\sqrt{i_{D1}} - \sqrt{i_{D2}} \right]$$

$$dv_{DI} = \sqrt{\frac{2}{\beta}} \left[\frac{1}{2} \frac{1}{\sqrt{i_{D1}}} di_{D1} - \frac{1}{2} \frac{1}{\sqrt{i_{D2}}} \frac{\partial i_{D2}}{\partial i_{D1}} di_{D1} \right]$$

$$i_{D2} = I_{SS} - i_{D1}$$

$$\frac{\partial i_{D2}}{\partial i_{D1}} = -1$$

$$\left. \frac{di_{D1}}{dv_{DI}} \right|_{v_{DI}=0} = \sqrt{\frac{\beta}{2}} \frac{1}{\frac{1}{2} \frac{1}{\sqrt{\frac{I_{SS}}{2}}} + \frac{1}{2} \frac{1}{\sqrt{\frac{I_{SS}}{2}}}}$$

$$= \sqrt{\frac{\beta}{2}} \sqrt{\frac{I_{SS}}{2}} = \frac{\sqrt{\beta I_{SS}}}{2}$$

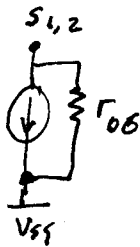
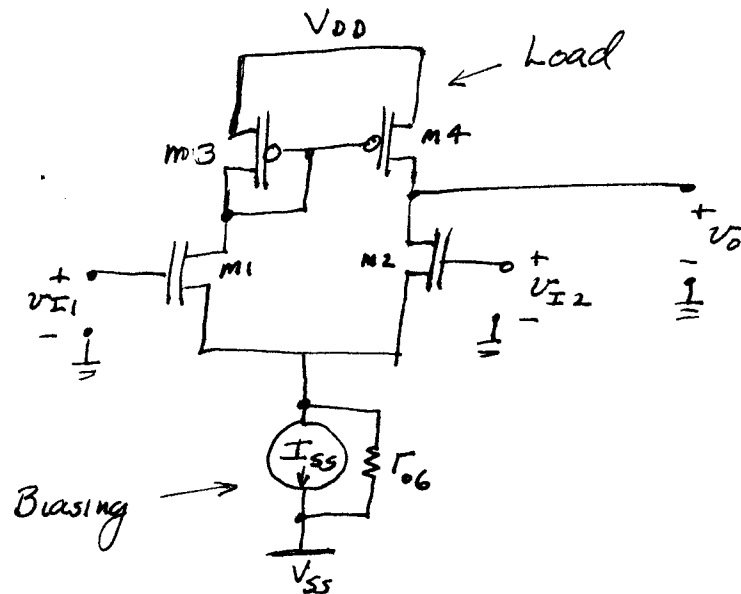
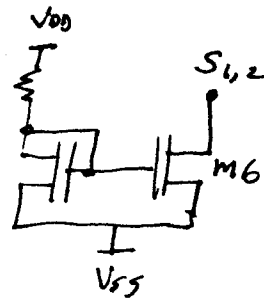
$$= \frac{1}{2} \sqrt{2\beta \underbrace{(I_{SS}/2)}_{I_{D1}}} = \frac{g_m}{2}$$

$$I_D = \frac{\beta}{2} (V_{GS} - V_t)^2$$

$$\frac{\partial I_D}{\partial V_{GS}} = \beta (V_{GS} - V_t) = \sqrt{I_D \beta 2}$$

Compare (24.12)

CURRENT SOURCE LOAD (24.1.1)


 $\Rightarrow$
Common Mode Range (CMR)

$$\text{SET } v_{I1} = v_{I2} = v_I$$

- As $v_I \rightarrow V_{SS}$, $V_{DS6} < (V_{GS6} - V_{THN})$
 $m6 \rightarrow \text{triode}.$

$$v_{I\text{MIN}} - V_{GS1} - V_{DS6} - V_{SS} = 0$$

$$v_{I\text{MIN}} = \sqrt{\frac{I_{SS}}{\beta_1}} + V_{THN} + \sqrt{\frac{2I_{SS}}{\beta_6}} + V_{SS}$$

W6

4

$$\frac{V_{I_{max}}}{V_{I_{max}}} =$$

• As $V_I \rightarrow V_{DD}$, $M1$ & $M2 \rightarrow$ triode

$$V_{DS1} = V_{GS1} - V_{THN}$$

$$V_{D1} = V_{G1} - V_{THN}$$

$$V_{I_{max}} = V_{G1} = V_{D1} + V_{THN}$$

$$V_{I_{max}} = V_{DD} - V_{SG3} + V_{THN}$$

~~$$V_{D1} + V_{SG1} + V$$~~

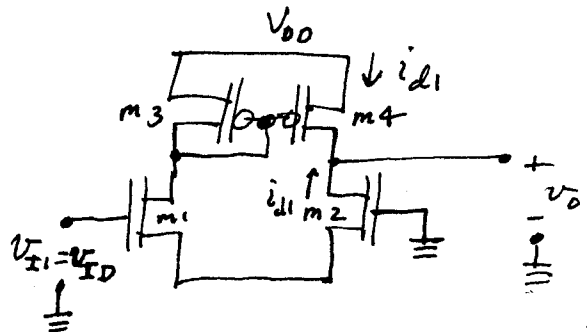
$$V_{D1} = V_{DD} - V_{SG3}$$

$$V_{I_{max}} = V_{DD} - \underbrace{\left[\sqrt{\frac{I_{SS}}{\beta_3}} + V_{THP} \right]}_{V_{SG3}} + V_{THN}$$

$$V_{I_{max}} \approx V_{DD} - \sqrt{\frac{I_{SS}}{\beta_3}}$$

SMALL SIGNAL GAIN

• Ground G2



$$V_{I1} = V_{ID} = V_{gs1} - V_{gs2} = \frac{i_{d1}}{g_{m1}} - \frac{i_{d2}}{g_{m2}}$$

$$i_{d1} = -i_{d2} = i_d$$

$$V_{i1} = i_d \frac{2}{g_m} \Rightarrow i_d \frac{g_m}{2} V_{i1}$$

Total I flowing into output node is
 $2 i_d$

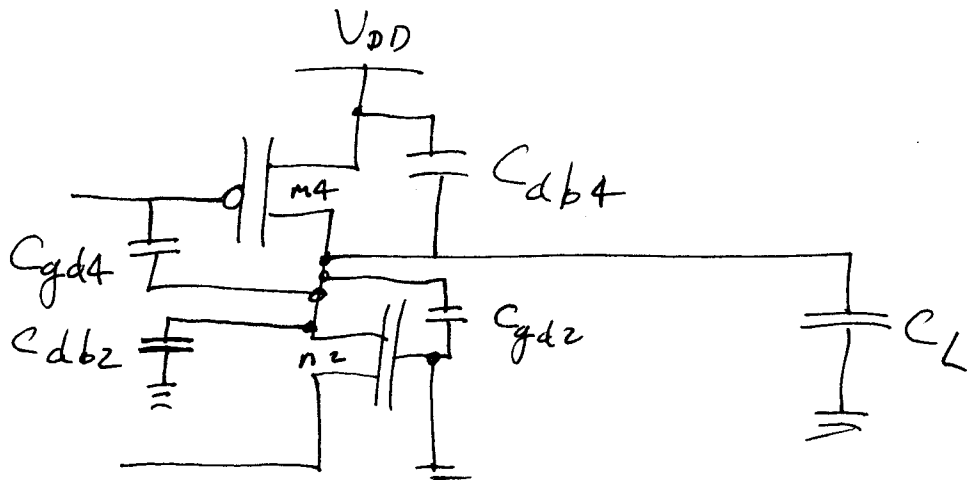
$$V_o = 2 i_d r_{o2} \parallel r_{o4} = g_m V_{i1} r_{o2} \parallel r_{o4}$$

$$A_{vd} = \frac{V_o}{V_{i1} - V_{i2}} = g_m r_{o2} \parallel r_{o4}$$

$$r_o = \frac{1}{\lambda I_{ss}/2} \quad g_m = \sqrt{\beta I_{ss}}$$

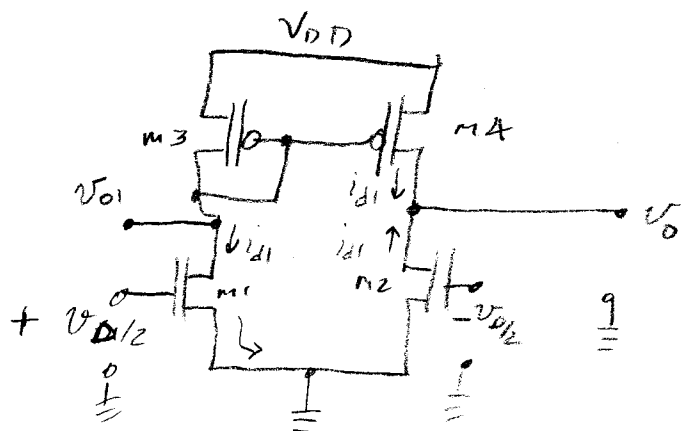
$$A_{vd} = \frac{\sqrt{\beta I_{ss}}}{(\lambda_2 + \lambda_4) \frac{I_{ss}}{2}} = \frac{2\sqrt{\beta}}{(\lambda_4 + \lambda_2) \sqrt{I_{ss}}}$$

SMALL SIGNAL FREQ. RESPONSE



$$R_o = r_{o2} \parallel r_{o4}$$

$$C_o = C_L + C_{db4} + C_{gd4} + C_{db2} + C_{gd2}$$

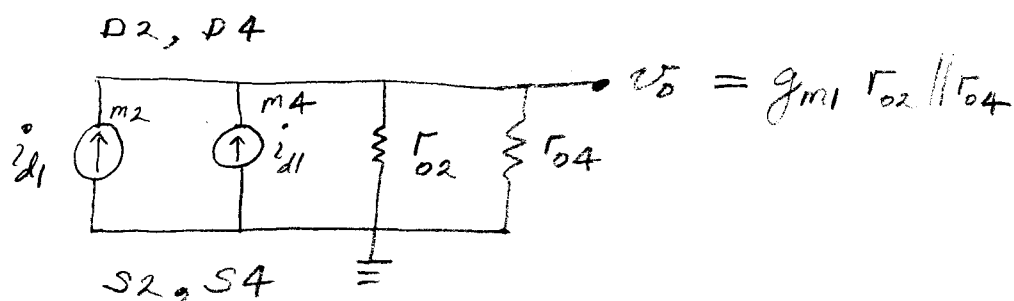


$$i_{D2} = -i_{D1}$$

$$\text{DC BIAS: } v_o = 0 \Rightarrow I_{D1} = I_{D2} = I_{SS}/2$$

$$g_m v_{gs2} = -g_m v_{gs1}$$

$\Rightarrow v_s$ is fixed at $V_s \Rightarrow$ a.c. ground



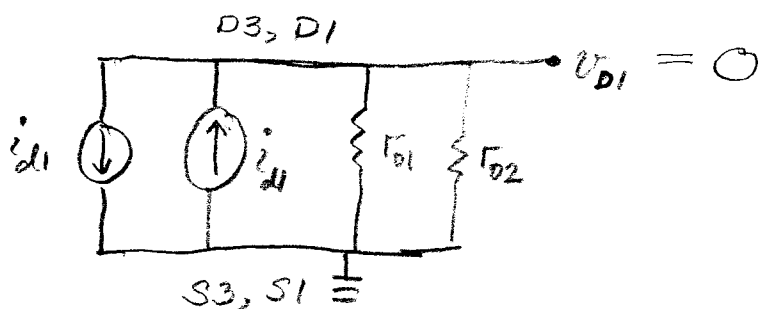
$$i_{D1} = g_{m1} v_{o/2}$$

$$v_o = g_{m1} \frac{v_o}{2} (2) (r_{O2} \parallel r_{O4})$$

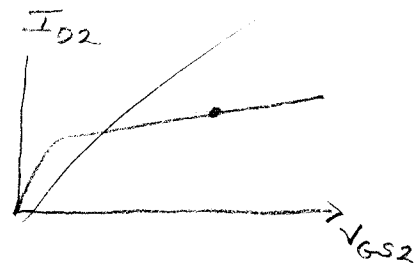
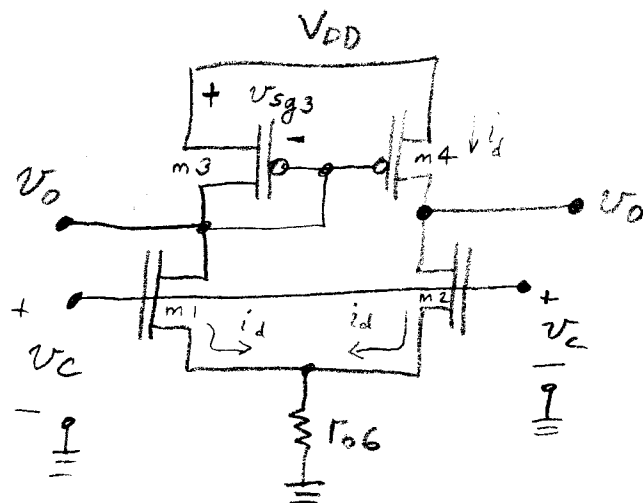
$$A_{v_D} \equiv \frac{v_o}{v_D} = g_{m1} (r_{O2} \parallel r_{O4})$$

(Differential Gain)

$$\approx \frac{v_o - v_{D1}}{v_D}$$



Common Mode Rejection Ratio (24.1.2) (CMRR)



Small-signal equivalent circuit for common-mode analysis (D2, 4):

$$\frac{v_o}{r_{o4}} + \frac{v_o - v_{s2}}{r_{o2}} = 0$$

$$- \frac{v_{s2}}{r_{o6}} + \frac{v_o - v_{s2}}{r_{o2}} + i_d = 0$$

$$v_{s2} \left(\frac{1}{r_{o6}} + \frac{1}{r_{o2}} \right) = \frac{v_o}{r_{o2}} + i_d$$

$$v_{s2} = \frac{v_o}{r_{o2}} \frac{r_{o6} r_{o2}}{r_{o6} + r_{o2}} + i_d \left(r_{o2} \parallel r_{o6} \right)$$

$$v_o \left(\frac{1}{r_{o4}} + \frac{1}{r_{o2}} \right) = v_o \frac{r_{o2} \parallel r_{o6}}{r_{o2}^2} + i_d \frac{r_{o2} \parallel r_{o6}}{r_{o2}}$$

$$v_o \left[\frac{1}{r_{o2} \parallel r_{o4}} - \frac{r_{o2} \parallel r_{o6}}{r_{o2}^2} \right] = i_d \frac{r_{o2} \parallel r_{o6}}{r_{o6}}$$

$$v_c = v_{gs1} + 2 i_d r_{o6}$$

$$v_c = v_{gs1} + 2i_d r_{o6}$$

$$= i_d \left(\frac{1}{g_{m1}} + 2r_{o6} \right)$$

$$v_o = -v_{sg3} = -\frac{i_d}{g_{m3}} = -\frac{i_d}{g_{m4}}$$

$$A_c \equiv \frac{v_o}{v_c} = \frac{-1/g_{m4}}{\frac{1}{g_{m1}} + 2r_{o6}}$$

$$A_c \approx \frac{-1}{2g_{m4} r_{o6}}$$

$$A_{v_D} = g_{m1} r_{o2} \parallel r_{o4}$$

$$CMRR = 20 \log_{10} \left| \frac{A_v}{A_c} \right| = 20 \log_{10} (2g_{m1} g_{m4} (r_{o2} \parallel r_{o4}) r_{o6})$$

$$\approx 40 \log_{10} (g_m r_o)$$

$r_{o6} = R_o$ of biasing I source

$$\boxed{As \ R_o \rightarrow \infty, \ A_c \rightarrow 0, \ CMRR \rightarrow \infty.}$$

For IDEAL CURRENT SOURCE, $CMRR = \infty$.

$$Z_{out} = C_0 r_{o2} \parallel r_{o4}$$

$$f_{3dB} = \frac{1}{2\pi Z_{out}} \approx \frac{1}{2\pi C_L r_{o2} \parallel r_{o4}}$$

SLEW RATE (Switching perspective)
Large signal

$$V_{I1} = \begin{array}{c} \text{---} \\ | \\ -V_{SS} \end{array} \Rightarrow \text{turns off } M_{1,3,4}$$

M_2 triode

All of I_{SS} drawn from C_L

$$\frac{dV}{dt} = \frac{I_{SS}}{C_L}$$

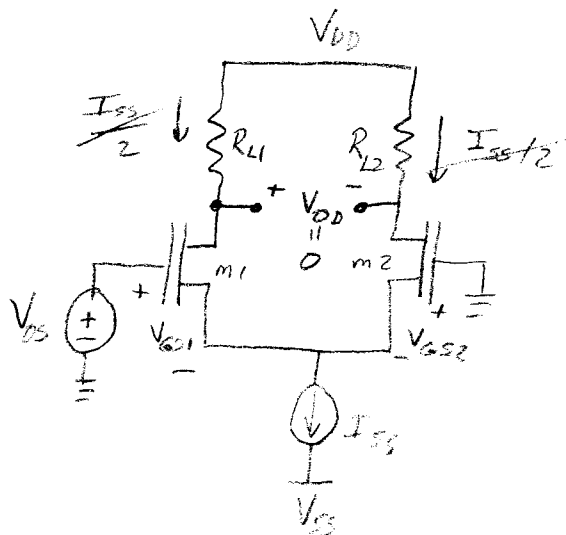
charge/
Max rate of discharge
(SLEW RATE)

$$V_{I1} = \begin{array}{c} V_{DD} \\ | \\ \text{---} \end{array} \Rightarrow M_{1,3,4} \rightarrow I_{SS} = I_D$$

All I_{SS} applied to C_L

$$\frac{dV}{dt} = \frac{I_{SS}}{C_L}$$

MATCHING



$$V_{OS} = V_{GS1} - V_{GS2} = V_{G1} - V_{G2} = V_{G1} = V_{OS}$$

$$V_{OS} = V_{t1} + \sqrt{\frac{2I_{D1}}{\beta_1}} - V_{t2} - \sqrt{\frac{2I_{D2}}{\beta_2}}$$

DEFINE :

$$\Delta I_D = I_{D1} - I_{D2}$$

$$I_D = (I_{D1} + I_{D2}) / 2$$

$$\Delta \left(\frac{W}{L}\right) = (W/L)_1 - (W/L)_2$$

$$\frac{W}{L} = [(W/L)_1 + (W/L)_2] / 2$$

$$\Delta V_t = V_{t1} - V_{t2}$$

$$V_t = (V_{t1} + V_{t2}) / 2$$

$$\Delta R_L = R_{L1} - R_{L2}$$

$$R_L = (R_{L1} + R_{L2}) / 2$$

$$I_{D1} R_{L1} = I_{D2} R_{L2}$$

$$I_{D2} = I_{D1} \frac{R_{L1}}{R_{L2}}$$

$$I_{D1} = \frac{\Delta I_D + 2I_D}{2}$$

$$V_{OS} = \Delta V_t + \frac{V_{GS} - V_t}{2} \left[\frac{-\Delta R_L}{R_L} - \frac{\Delta(W/L)}{W/L} \right]$$

~~W6.12~~

Effect of ΔR_L & $\Delta \frac{W}{L}$ reduced by $\downarrow (V_{GS} - V_t)$.
 $\Delta V_t \sim 2 \text{ mV} \sim \sigma$ over die (wafer).

Layout : Common Centroid layout. Fig. 24.15.

CR0